

**Numerical Methods of
Electromagnetic Field Theory II (NFT II)
Numerische Methoden der
Elektromagnetischen Feldtheorie II (NFT II) /**

5th Lecture / 5. Vorlesung

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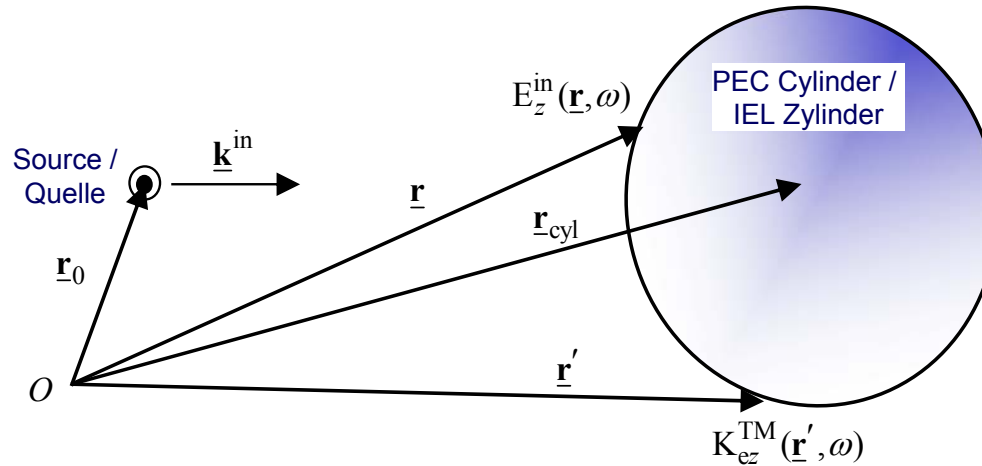
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EM Scattering by a Perfectly Electrically Conducting Cylinder: EFIE Discretized in the 2-D TM Case with Pulse Basis and Delta Testing Functions / EM-Streuung an einem ideal elektrisch leitendem Zylinder: EFIE diskretisiert im 2D-TM-Fall mit Impuls-Basisfunktionen und Delta-Testfunktionen



2-D Case /
2D-Fall

$$\mathbf{R} = \underbrace{r\mathbf{e}_r(\varphi)}_{=\mathbf{r}} + \underbrace{z\mathbf{e}_r(\varphi)}_{=0}$$

$$= \mathbf{r}$$

2-D PEC TM EFIE / 2D-IEL-TM-EFIE

$$-j\omega\mu_0 \oint_{\mathbf{r}' \in C_{sc} = \partial S_{sc}} \mathbf{K}_{ez}^{TM}(\mathbf{r}', \omega) G(\mathbf{r} - \mathbf{r}', \omega) d\mathbf{r}' = E_z^{in}(\mathbf{r}, \omega), \quad \mathbf{r} \in C_{sc}$$

This is a *Fredholm integral equation of the 1. kind* in form of a *closed line integral* for the *unknown* electric surface current density for a *known* incident field. /
Dies ist eine *Fredholmsche Integralgleichung 1. Art* in Form eines *geschlossenen Linienintegrals* für die *unbekannte* elektrische Flächenladungsdichte für ein *bekanntes* einfallendes Feld.

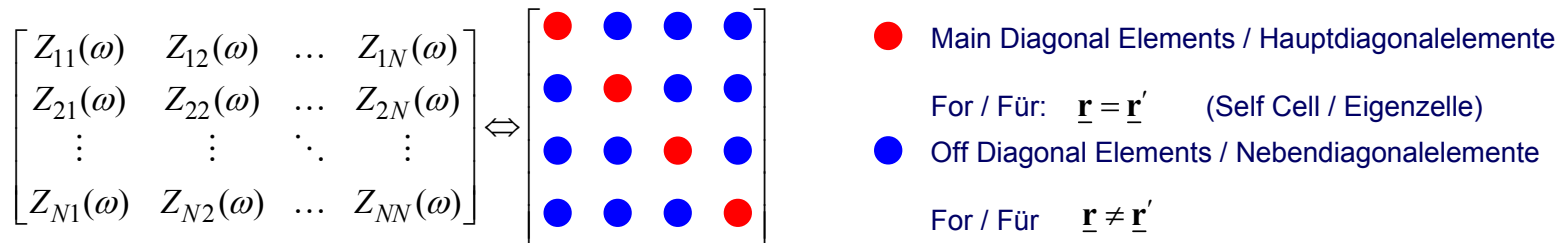
$$G(\mathbf{r} - \mathbf{r}', \omega) = \frac{j}{4} H_0^{(1)}(k_0 |\mathbf{r} - \mathbf{r}'|)$$

EM Scattering by a Perfectly Electrically Conducting Cylinder: EFIE Discretized in the 2-D TM Case with Pulse Basis and Delta Testing Functions / EM-Streuung an einem ideal elektrisch leitendem Zylinder: EFIE diskretisiert im 2D-TM-Fall mit Impuls-Basisfunktionen und Delta-Testfunktionen

2-D PEC TM EFIE / 2D-IEL-TM-EFIE

$$-j\omega\mu_0 \oint_{\mathbf{r}' \in C_{sc} = \partial S_{sc}} \mathbf{K}_{ez}^{\text{TM}}(\mathbf{r}', \omega) G(\mathbf{r} - \mathbf{r}', \omega) d\mathbf{r}' = E_z^{\text{in}}(\mathbf{r}, \omega), \quad \mathbf{r} \in C_{sc}$$

We have to Consider Two Different Cases for the Elements of the Impedance Matrix /
Man unterscheidet zwei Verschiedene Fälle für die Elemente der Impedanzmatrix



- **Main Diagonal Elements / Hauptdiagonalelemente**
 1. Flat Cell Approximation / Ebene-Zelle-Approximation
 2. Power Series Expansion of the Hankel Function for Small Arguments / Potenzreihen-Approximation der Hankel-Funktion für kleine Argumente
- **Off Diagonal Elements / Nebendiagonalelemente**
 1. Flat Cell Approximation / Ebene-Zelle-Approximation
 2. Application of the Midpoint Rule / Anwendung der Mittelpunktsregel

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Elements of the Impedance Matrix /
Elemente der Impedanzmatrix

$$Z_{mn}(\omega) = \frac{\omega\mu_0}{4} \Delta^{(n)} \begin{cases} 1 + j\frac{2}{\pi} \left[\ln\left(\frac{k}{4} \Delta^{(n)}\right) + \gamma - 1 \right] & m = n \\ H_0^{(1)}(k r_{mn}) & m \neq n \end{cases}$$

Matrix Equation / Matrixgleichung

$$\underbrace{[Z]}_{=V/A}(\omega) \underbrace{\{K_{ez}^{TM}\}}_{=A/m}(\omega) = \underbrace{\{E_z^{in}\}}_{=V/m}(\omega)$$

Problem: Large Impedance Matrix /
Problem: Große Impedanzmatrix !



Iterative Solution via Conjugate Gradient (CG) Method /
Iterative Lösung durch Konjugierte Gradienten (KG) Methode

Solution of the Matrix Equation / Lösung der Matrixgleichung

$$\underbrace{\{K_{ez}^{TM}\}}_{=A/m}(\omega) = \underbrace{[Z]^{-1}}_{=A/V}(\omega) \underbrace{\{E_z^{in}\}}_{=V/m}(\omega)$$

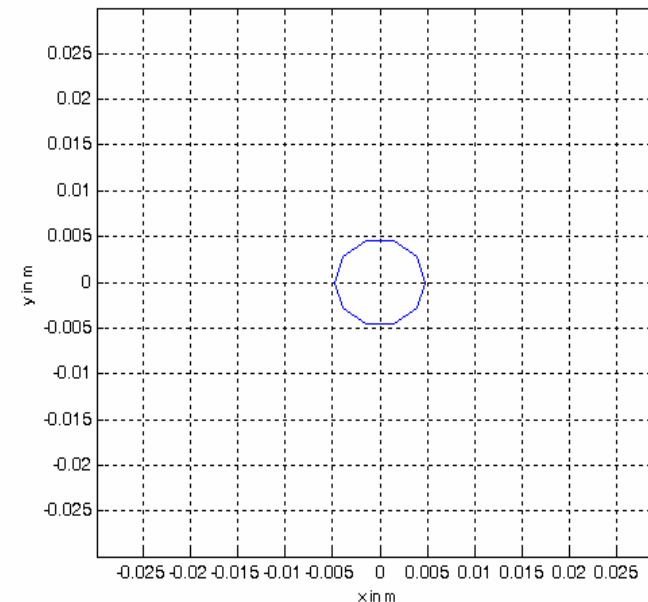
MATLAB Implementation / MATLAB-Implementierung

MATLAB Program to Generate the Geometry of
a Circular Cylinder /
MATLAB-Programm zur Generierung der Geometrie eines
kreisförmigen Zylinders

$$\underline{\mathbf{R}} = a \cos \varphi \underline{\mathbf{e}}_x + a \sin \varphi \underline{\mathbf{e}}_y \quad 0 \leq \varphi \leq 2\pi$$

```
sca_grid.nodes = zeros(N+1,3);  
sca_grid.number_of_nodes = N;  
for j=1:N+1  
    phi_m = 2.0*M_PI*real((j-1))/real(N);  
    sca_grid.nodes(j,1) = a * cos( phi_m ); % x component  
    sca_grid.nodes(j,2) = a * sin( phi_m ); % y component  
    sca_grid.nodes(j,3) = 0; % z component  
end  
circumference = 2.0*M_PI*a;
```

Geometry of a Circular Cylinder /
Geometrie des kreisförmigen Zylinders



EM Scattering by a Perfectly Electrically Conducting Cylinder: EFIE Discretized in the 2-D TM Case with Pulse Basis and Delta Testing Functions / EM-Streuung an einem ideal elektrisch leitendem Zylinder: EFIE diskretisiert im 2D-TM-Fall mit Impuls-Basisfunktionen und Delta-Testfunktionen

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Elemente der Impedanzmatrix

$$Z_{mn}(\omega) = \frac{\omega\mu_0}{4} \Delta^{(n)} \begin{cases} 1 + j\frac{2}{\pi} \left[\ln\left(\frac{k}{4} \Delta^{(n)}\right) + \gamma - 1 \right] & m = n \\ H_0^{(1)}(k r_{mn}) & m \neq n \end{cases}$$

MATLAB Program to Compute the Impedance Matrix / MATLAB-Programm zur Berechnung der Impedanzmatrix

```

for j=1:N % ___ loop for r_m the observation point ___ */

%/* ___ Coordinates of the observation point r_m ___ */
vrn(1) = ( sca_grid.nodes(j,1) + sca_grid.nodes(j+1,1) )/2;
vrn(2) = ( sca_grid.nodes(j,2) + sca_grid.nodes(j+1,2) )/2;
vrn(3) = ( sca_grid.nodes(j,3) + sca_grid.nodes(j+1,3) )/2;

for i=1:N

%/* ___ vrpn is the phase center of the ith element ___ */
vrn(1) = ( sca_grid.nodes(i,1) + sca_grid.nodes(i+1,1) )/2;
vrn(2) = ( sca_grid.nodes(i,2) + sca_grid.nodes(i+1,2) )/2;
vrn(3) = ( sca_grid.nodes(i,3) + sca_grid.nodes(i+1,3) )/2;

%/* ___ Difference vector vd of the ith element ___ */
vd(1) = sca_grid.nodes(i+1,1) - sca_grid.nodes(i,1);
vd(2) = sca_grid.nodes(i+1,2) - sca_grid.nodes(i,2);
vd(3) = sca_grid.nodes(i+1,3) - sca_grid.nodes(i,3);

Delta = norm(vd);

if j == i
    Z(i,j) = 0.25*omega*mu0*Delta
        * complex(1.0,2.0/M_PI *( log(0.25*k*Delta) + M_GAMMA-1.0
        ));
else %/* ___ Calculate off-diagonal ___ */
    vrnm(1) = vrm(1) - vrn(1);
    vrnm(2) = vrm(2) - vrn(2);
    vrnm(3) = vrm(3) - vrn(3);

    rmn = norm(vrnm);

%/* ___ Calculate Hankel function: H^1_0(z) ___ */
k_rmn = k * rmn;
z      = complex( k_rmn, 0); %/* ___ Complex argument ___ */

nu  = 0; %/* ___ initial order: n=0 ___ */
kind = 1; %/* ___ compute 1st kind ___ */

[H10, ierr] = besselh(nu,kind,z);

Z(i,j) = 0.25 * omega * mu0 * Delta * H10;

end
end
end

```

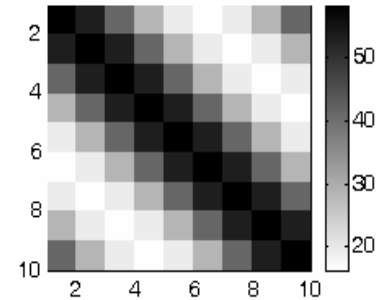
EM Scattering by a Perfectly Electrically Conducting Cylinder: EFIE Discretized in the 2-D TM Case with Pulse Basis and Delta Testing Functions / EM-Streuung an einem ideal elektrisch leitendem Zylinder: EFIE diskretisiert im 2D-TM-Fall mit Impuls-Basisfunktionen und Delta-Testfunktionen

Elements of the Impedance Matrix /
Elemente der Impedanzmatrix

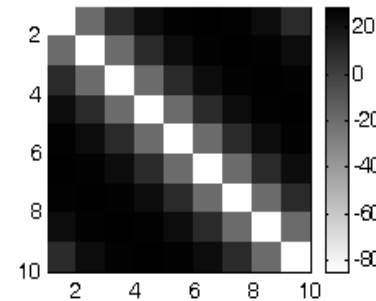
$$Z_{mn}(\omega) = \frac{\omega\mu_0}{4} \Delta^{(n)} \begin{cases} 1 + j\frac{2}{\pi} \left[\ln\left(\frac{k}{4} \Delta^{(n)}\right) + \gamma - 1 \right] & m = n \\ H_0^{(1)}(kr_{mn}) & m \neq n \end{cases}$$

$$ka = 1, \quad N = 10$$

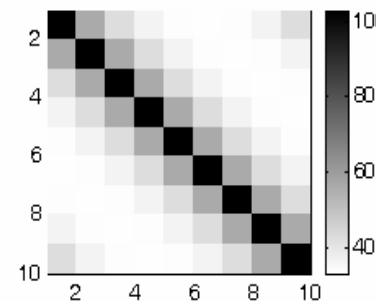
$\text{Re}\{[\mathbf{Z}](\omega)\}$



$\text{Im}\{[\mathbf{Z}](\omega)\}$



$[\mathbf{Z}](\omega)$



Iterative Methods for the Solution of Discrete Integral Equations / Iterative Methode zur Lösung von diskreten Integralgleichungen

CG Method – Conjugate Gradient (CG) Method

M. R. Hestenes & E. Stiefel, 1952

BiCG Method – Biconjugate Gradient (BiCG) Method

C. Lanczos, 1952
D. A. H. Jacobs, 1981
C. F. Smith et al., 1990
R. Barret et al., 1994

CGS Method – Conjugate Gradient Squared (CGS) Method (MATLAB Function)

P. Sonneveld, 1989

GMRES Method – Generalized Minimal – Residual (GMRES) Method

Y. Saad & M. H. Schultz, 1986
R. Barret et al., 1994
Y. Saad, 1996

QMR Method – Quasi–Minimal–Residual (QMR) Method

R. Freund & N. Nachtigal, 1990
N. Nachtigal, 1991
R. Barret et al., 1994
Y. Saad, 1996

MATLAB Function CGS – Conjugate Gradient Squared / MATLAB-Funktion CGS – Konjugierte Gradienten Quadriert

cgs Conjugate Gradients Squared method Syntaxx =
cgs(A,b)

cgs(A,b,tol)

cgs(A,b,tol,maxit)

cgs(A,b,tol,maxit,M)

cgs(A,b,tol,maxit,M1,M2)

cgs(A,b,tol,maxit,M1,M2,x0)

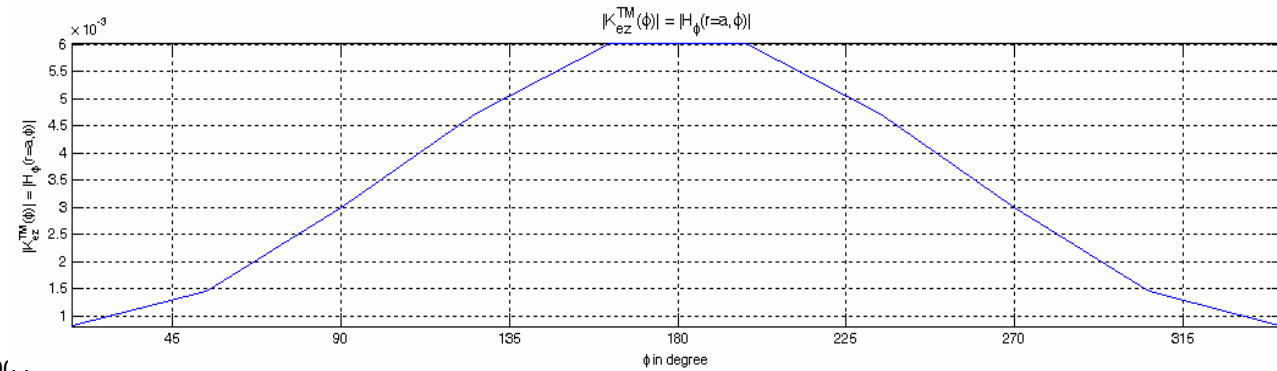
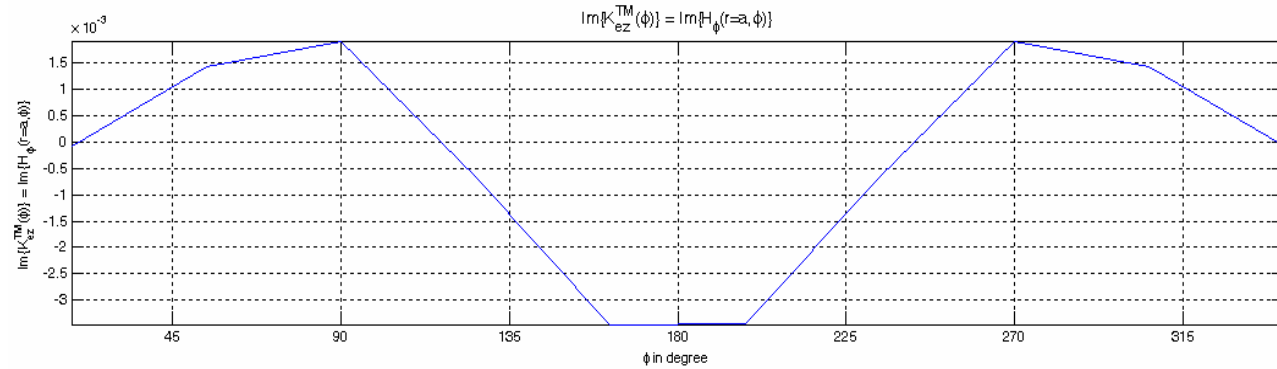
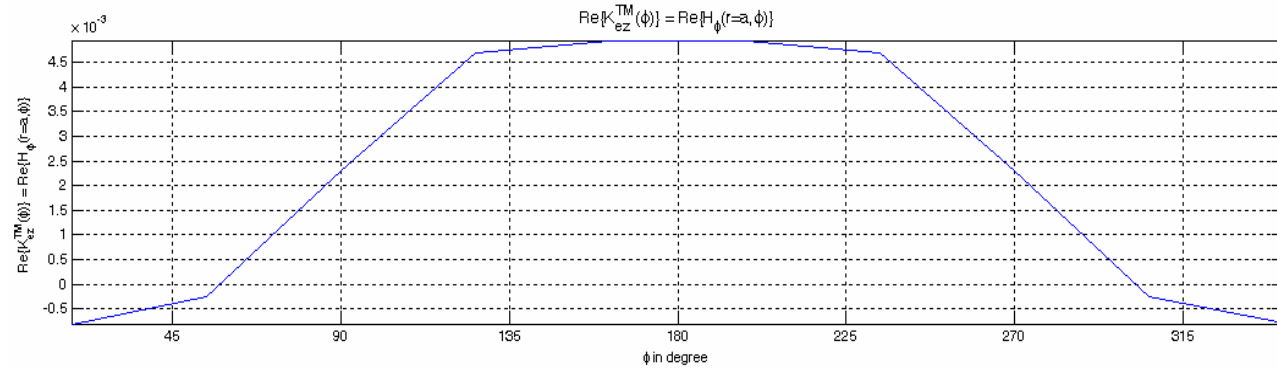
cgs(afun,b,tol,maxit,m1fun,m2fun,x0,p1,p2,...)

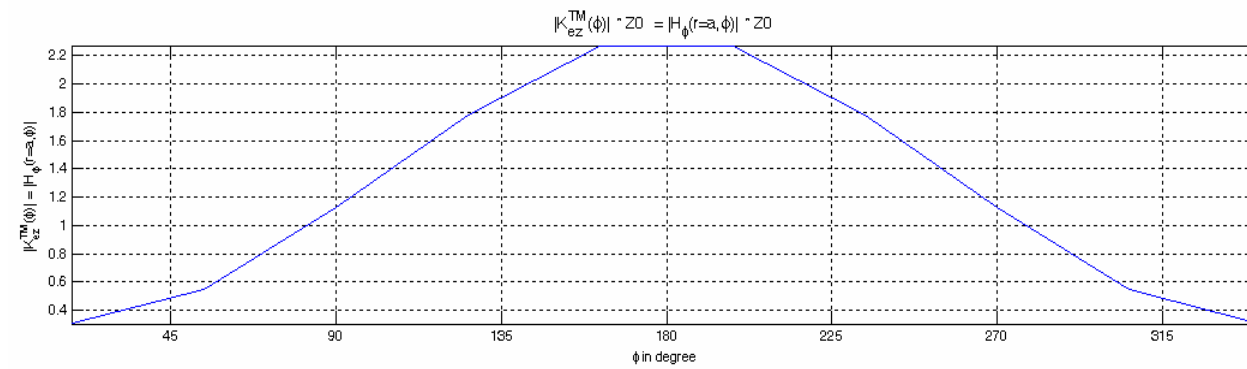
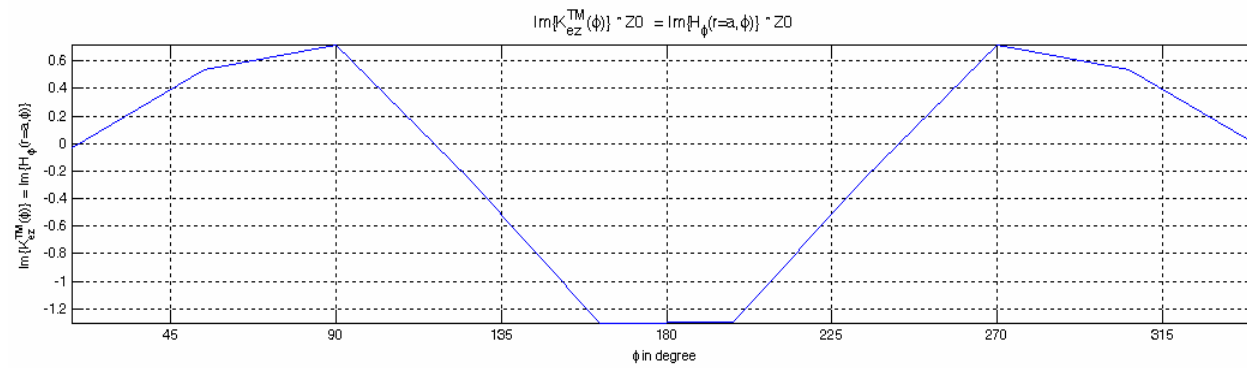
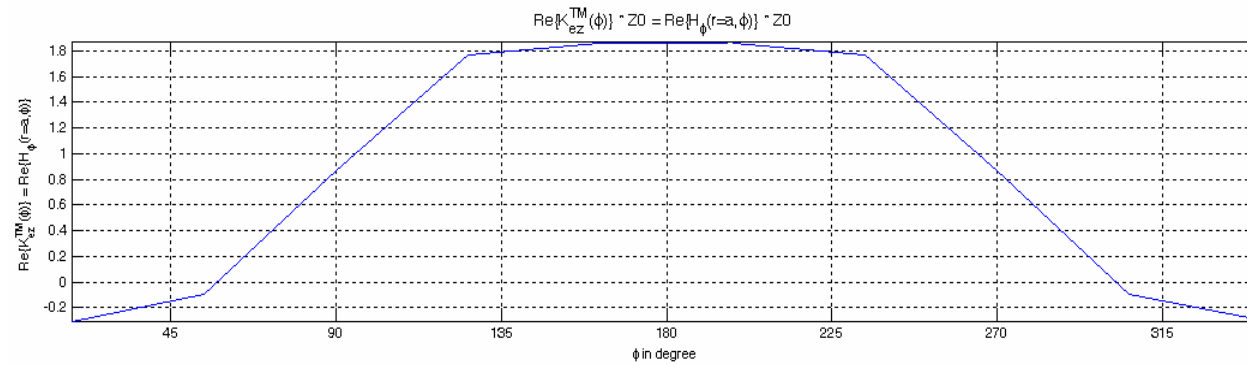
[x,flag] = cgs(A,b,...)

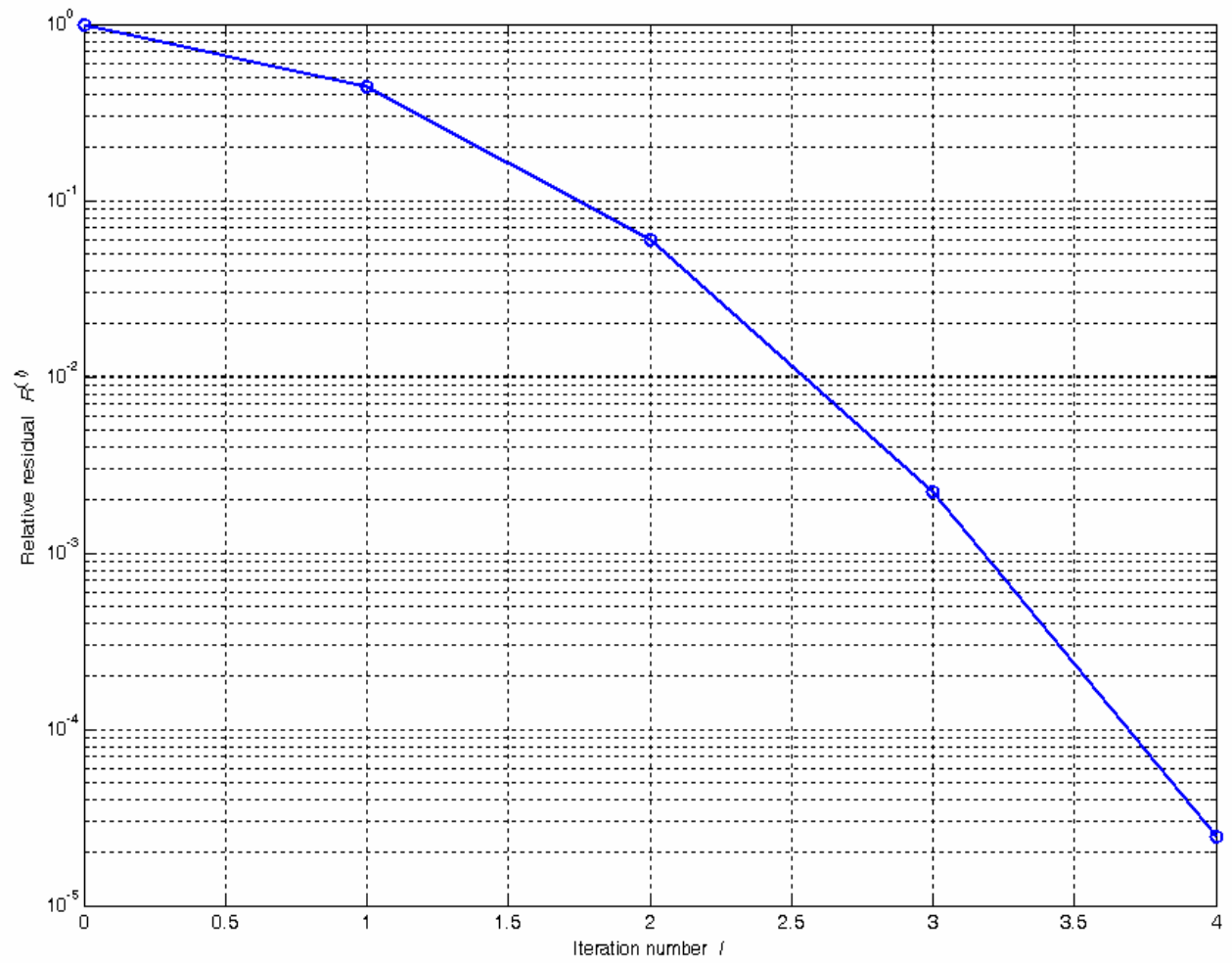
[x,flag,relres] = cgs(A,b,...)

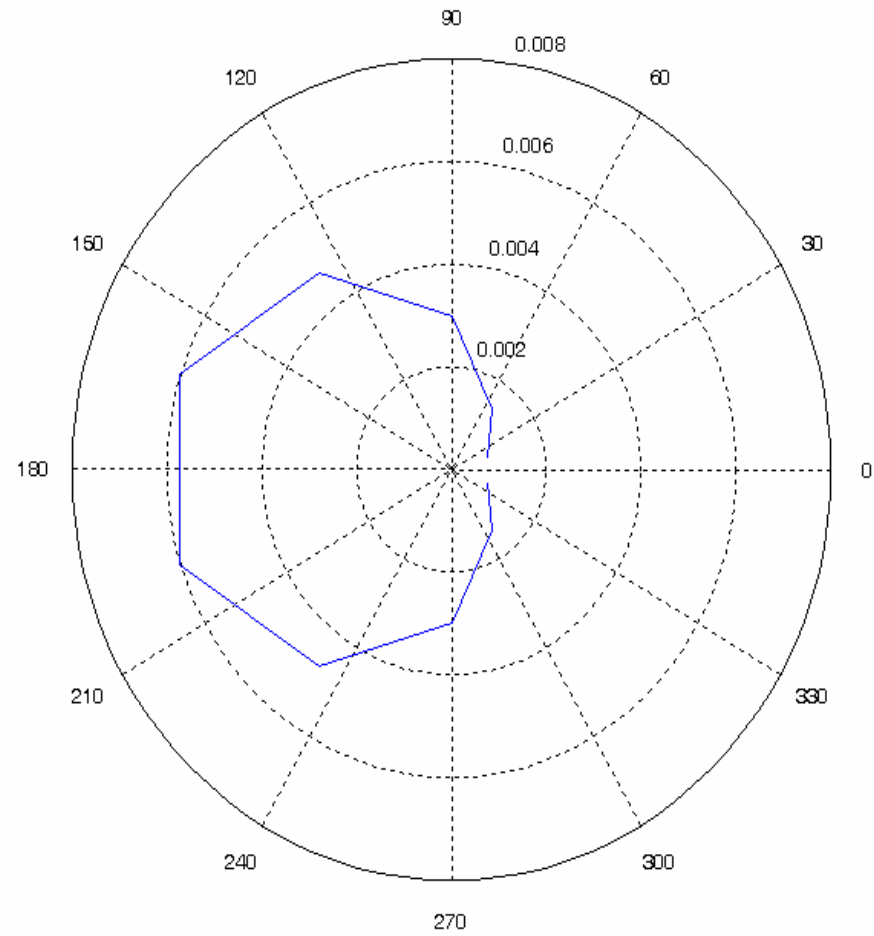
[x,flag,relres,iter] = cgs(A,b,...)

[x,flag,relres,iter,resvec] = cgs(A,b,...)





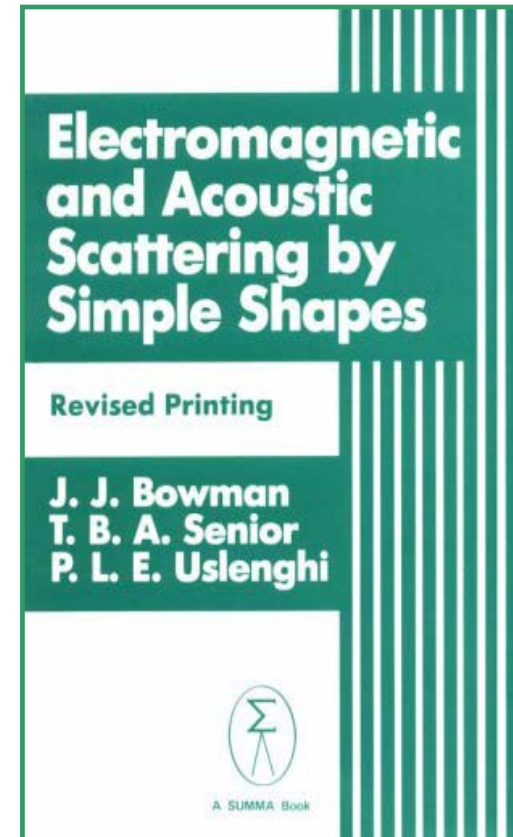
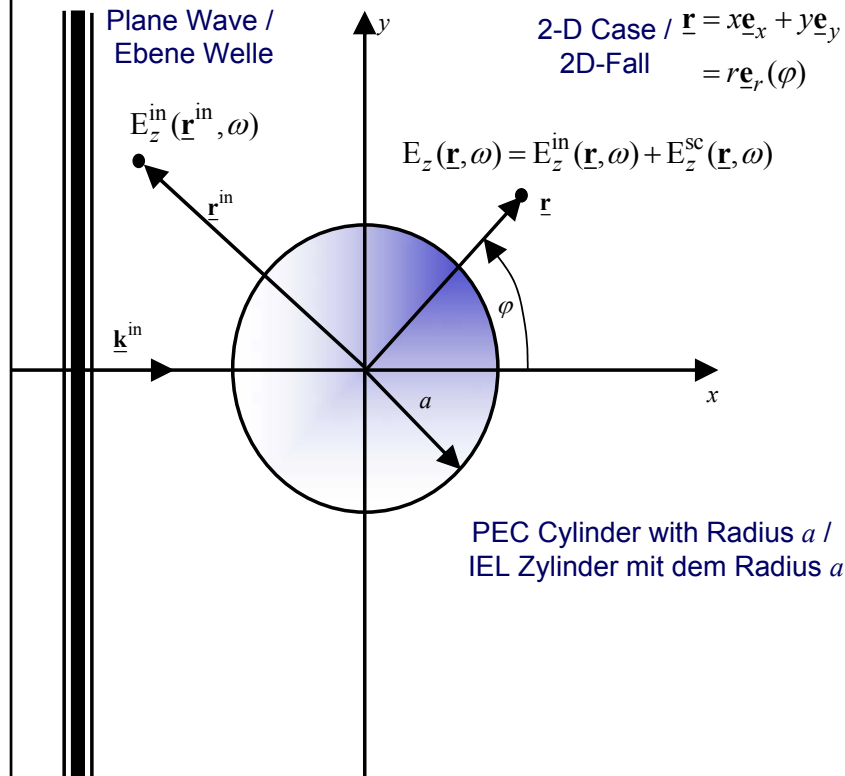




Diffraction of an EM Plane Wave on a Circular PEC Cylinder – TM Case / Beugung einer EM Ebenen Welle an einem kreisrunden IEL-Zylinder – TM-Fall

Separation of Variables
Analytic Solution in Terms of Eigenfunctions /
Separation der Variablen
Analytische Lösung in Form von Eigenfunktionen

J. J. Bowman, T. B. A. Senior, P. L. E. Uslenghi (Editors):
Electromagnetic and Acoustic Scattering by Simple Shapes.
Taylor & Francis Inc, New York, 1988.



Diffraction of an EM Plane Wave on a Circular PEC Cylinder – TM Case – Analytic Solution: Separation of Variables / Beugung einer EM Ebenen Welle an einem kreisrunden IEL-Zylinder – TM-Fall – Analytische Lösung: Separation der Variablen

Electric Field Strength of the Incident Wave /
Elektrische Feldstärke der einfallenden Welle

$$E_z^{\text{in}}(r, \varphi, \varphi_{\text{in}}, \omega) = \underbrace{E_0(\omega)}_{=1 \text{ V/m}} e^{j\mathbf{k}^{\text{in}} \cdot \mathbf{r}}$$

Boundary Condition at the PEC Cylinder /
Randbedingung am IEL-Zylinder

$$E_z(r = a, \varphi, \varphi_{\text{in}}, \omega) = E_z^{\text{in}}(r = a, \varphi, \varphi_{\text{in}}, \omega) + E_z^{\text{sca}}(r = a, \varphi, \varphi_{\text{in}}, \omega) = 0$$

Solution /
Lösung



Electric Field Strength of the Scattered Wave /
Elektrische Feldstärke der gestreuten Welle

$$E_z^{\text{sc}}(r, \varphi, \varphi_{\text{in}}, \omega) = - \sum_{n=0}^{\infty} \varepsilon_n (-j)^n \frac{J_n(ka)}{H_n^{(1)}(ka)} H_n^{(1)}(kr) \cos[n(\varphi - \varphi_{\text{in}})]$$

Neumann Function /
Neumann-Funktion

$$\varepsilon_n = \begin{cases} 1 & n = 0 \\ 2 & n = 1, 2, 3, \dots \end{cases}$$

Diffraction of an EM Plane Wave on a Circular PEC Cylinder – TM Case – Analytic Solution: Separation of Variables / Beugung einer EM Ebenen Welle an einem kreisrunden IEL-Zylinder – TM-Fall – Analytische Lösung: Separation der Variablen

Boundary Condition at the PEC Cylinder /
Randbedingung am IEL-Zylinder

$$\underline{\mathbf{n}} \times \underline{\mathbf{E}}(r = a, \varphi, \varphi_{\text{in}}, \omega) = \underline{\mathbf{0}}$$

$$E_z(r = a, \varphi, \varphi_{\text{in}}, \omega) = 0$$

Induced Electric Surface Current Density at /
Induzierte elektrische Flächenstromdichte bei $r = a$

$$\underline{\mathbf{n}} \times \underline{\mathbf{H}}(r = a, \varphi, \varphi_{\text{in}}, \omega) = \underline{\mathbf{K}}_e(r = a, \varphi, \varphi_{\text{in}}, \omega), \quad \underline{\mathbf{n}} = \underline{\mathbf{e}}_R$$

$$\begin{aligned} K_{ez}^{\text{TM}}(\varphi, \varphi_{\text{in}}, \omega) &= H_\varphi(r = a, \varphi, \varphi_{\text{in}}, \omega) \\ &= H_\varphi^{\text{in}}(r = a, \varphi, \varphi_{\text{in}}, \omega) + H_\varphi^{\text{sc}}(r = a, \varphi, \varphi_{\text{in}}, \omega) \\ &= 2 \frac{Y_0}{\pi} \frac{1}{ka} \sum_{n=0}^{\infty} \varepsilon_n \frac{(-j)^n}{H_n^{(1)}(ka)} \cos[n(\varphi - \varphi_{\text{in}})] \end{aligned}$$

$$K_{ez}^{\text{TM}}(\varphi, \varphi_{\text{in}}, \omega) = 2 \frac{Y_0}{\pi} \frac{1}{ka} \sum_{n=0}^{\infty} \varepsilon_n \frac{(-j)^n}{H_n^{(1)}(ka)} \cos[n(\varphi - \varphi_{\text{in}})]$$

MATLAB Programme / MATLAB-Programm

MATLAB Program / MATLAB-Programm

```
for nka=1:N_max_kas

ka = max_kas(nka); % max_kas = {1, 5, 10}

a = ka / k;

legend_matrix(nka,:) = sprintf('ka = %2d',ka)

for nphi=1:Nphi

phi(nphi)      = (nphi-1)*2.0*pi/(Nphi-1);
phi_deg(nphi)  = (nphi-1)*2.0*pi/(Nphi-1)*180.0/pi;

Hphi(nphi,nka) = 0.0;

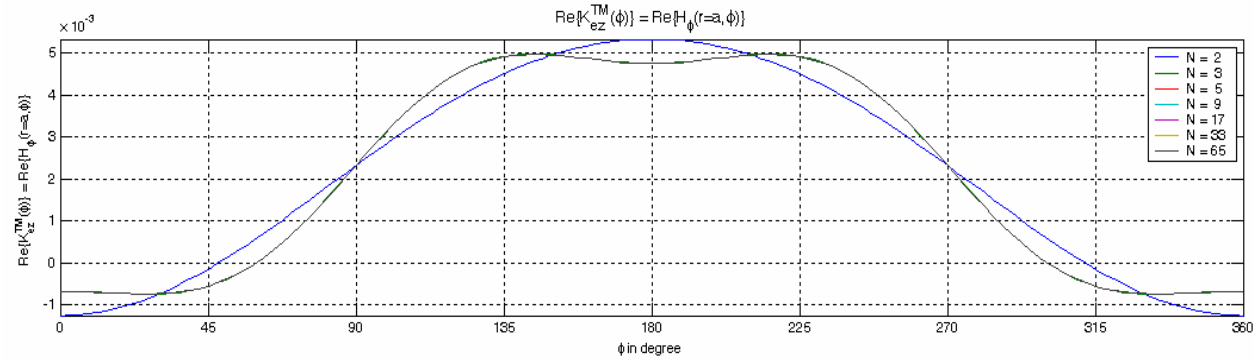
for n = 0:N
Hphi(nphi,nka) = Hphi(nphi,nka) + epsilon_n(n+1) * (complex(0,-1))^n / besselh(n,1,ka) * cos(n*(phi(nphi)-phi_in));
end

% Magnetic field strength component / Magnetische Feldstärkekomponente
Hphi(nphi,nka) = Hphi(nphi,nka) * 2.0/M_PI * Y0/ka;
% Normalized magnetic field strength component / Normierte magnetische Feldstärkekomponente
Hphi_Z0(nphi,nka) = Hphi(nphi,nka) * Z0;

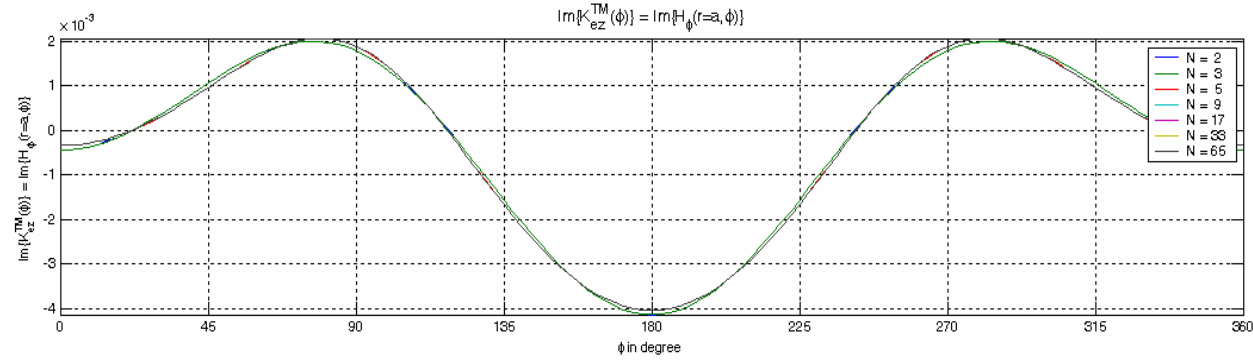
end
end
```

Induced Electric Surface Current Density for Different Order N , $ka = 1$ / Induzierte elektrische Flächenstromdichte für verschiedene Ordnungen N , $ka = 1$

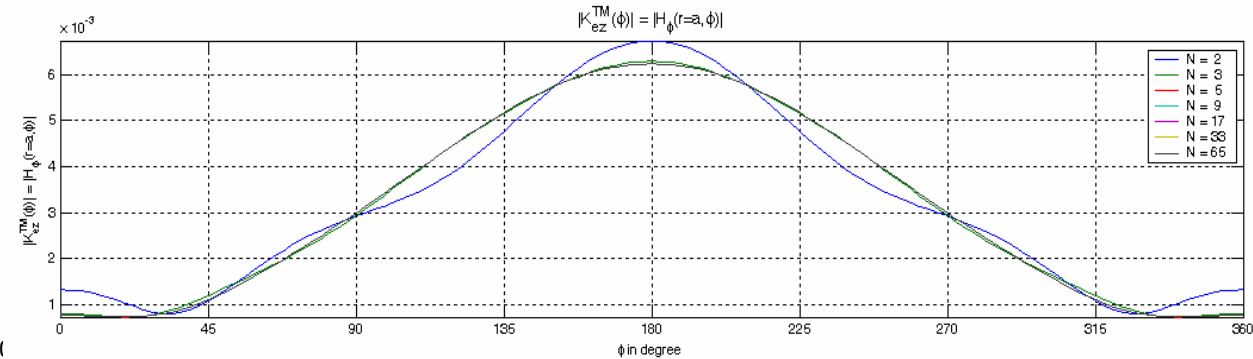
$$\operatorname{Re}\{K_{ez}^{\text{TM}}(\varphi, \varphi_{\text{in}} = 0^\circ, \omega)\}$$



$$\operatorname{Im}\{K_{ez}^{\text{TM}}(\varphi, \varphi_{\text{in}} = 0^\circ, \omega)\}$$

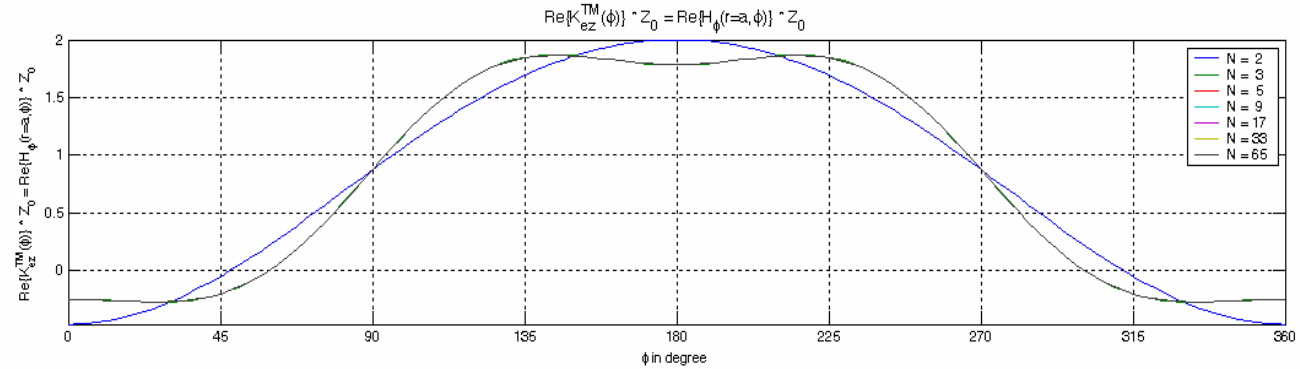


$$|K_{ez}^{\text{TM}}(\varphi, \varphi_{\text{in}} = 0^\circ, \omega)|$$

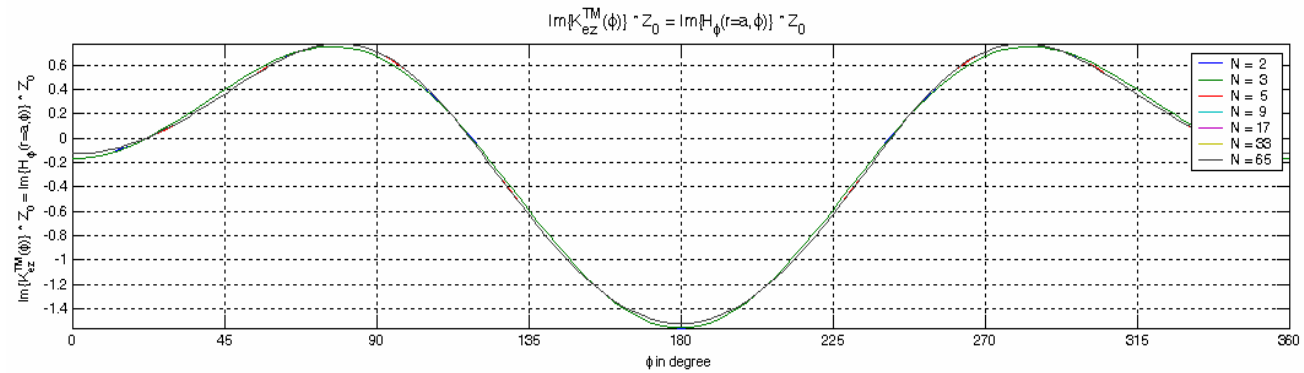


Induced Electric Surface Current for Different Order N , $ka = 1$ / Induzierte elektrische Flächenstrom für verschiedene Ordnungen N , $ka = 1$

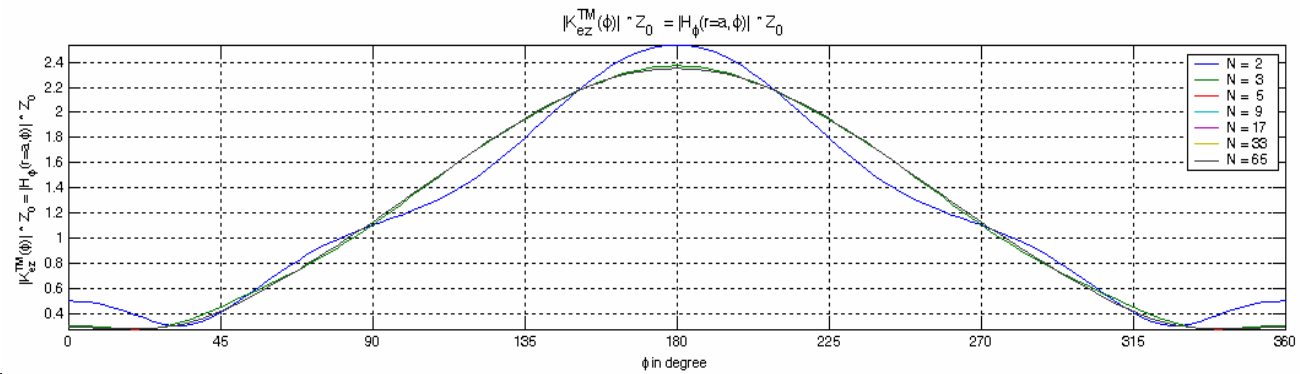
$$Z_0 \operatorname{Re}\{K_{ez}^{\text{TM}}(\varphi, \varphi_{\text{in}} = 0^\circ, \omega)\}$$



$$Z_0 \operatorname{Im}\{K_{ez}^{\text{TM}}(\varphi, \varphi_{\text{in}} = 0^\circ, \omega)\}$$

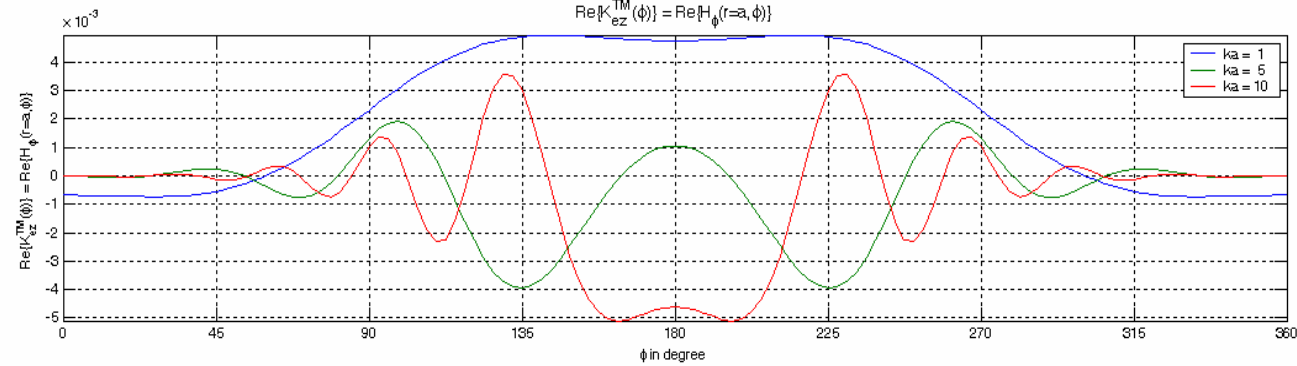


$$Z_0 |K_{ez}^{\text{TM}}(\varphi, \varphi_{\text{in}} = 0^\circ, \omega)|$$

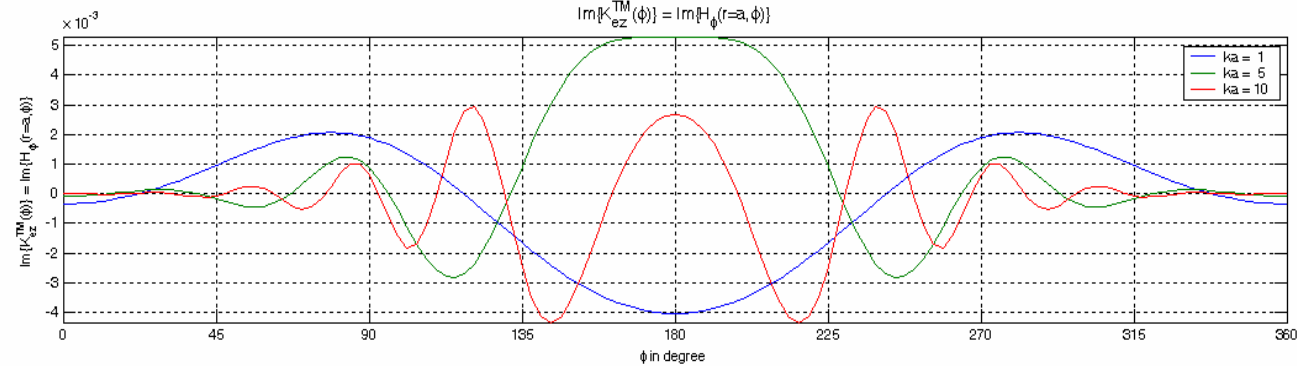


Induced Electric Surface Current for Different $ka = \{1, 5, 10\}$ and $N = 128$ Induzierter elektrischer Flächenstrom für verschiedene $ka = \{1, 5, 10\}$ und $N = 128$

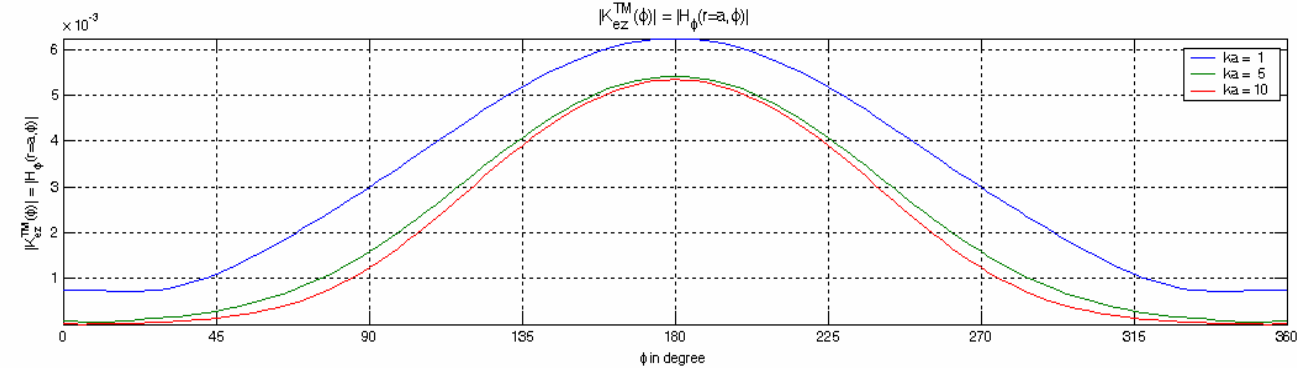
$$\operatorname{Re}\{K_{ez}^{\text{TM}}(\varphi, \varphi_{\text{in}} = 0^\circ, \omega)\}$$



$$\operatorname{Im}\{K_{ez}^{\text{TM}}(\varphi, \varphi_{\text{in}} = 0^\circ, \omega)\}$$

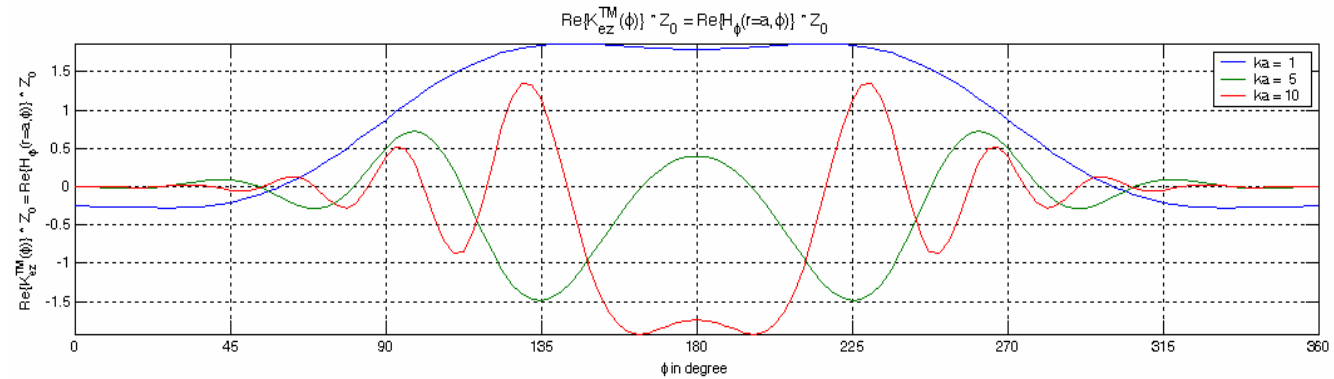


$$|K_{ez}^{\text{TM}}(\varphi, \varphi_{\text{in}} = 0^\circ, \omega)|$$

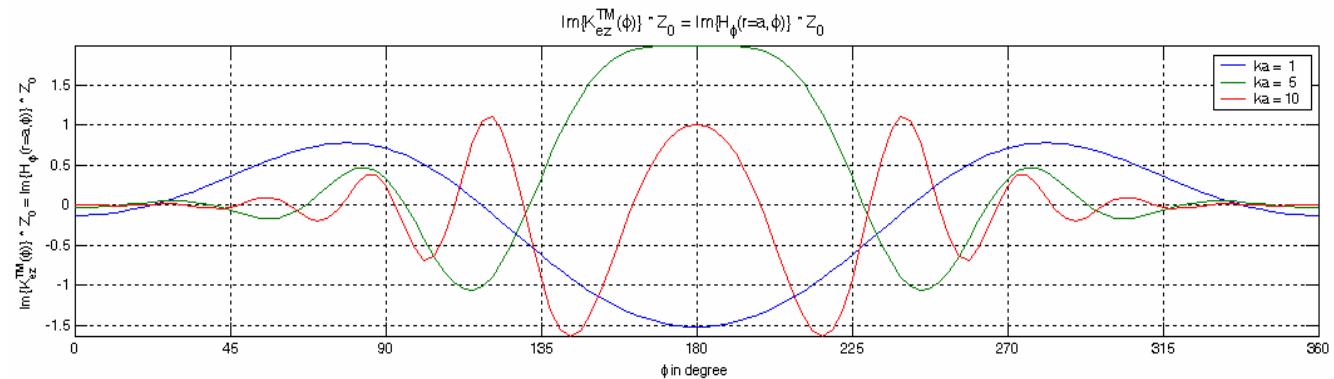


Induced Electric Surface Current for Different $ka = \{1, 5, 10\}$ and $N = 128$ Induzierter elektrischer Flächenstrom für verschiedene $ka = \{1, 5, 10\}$ und $N = 128$

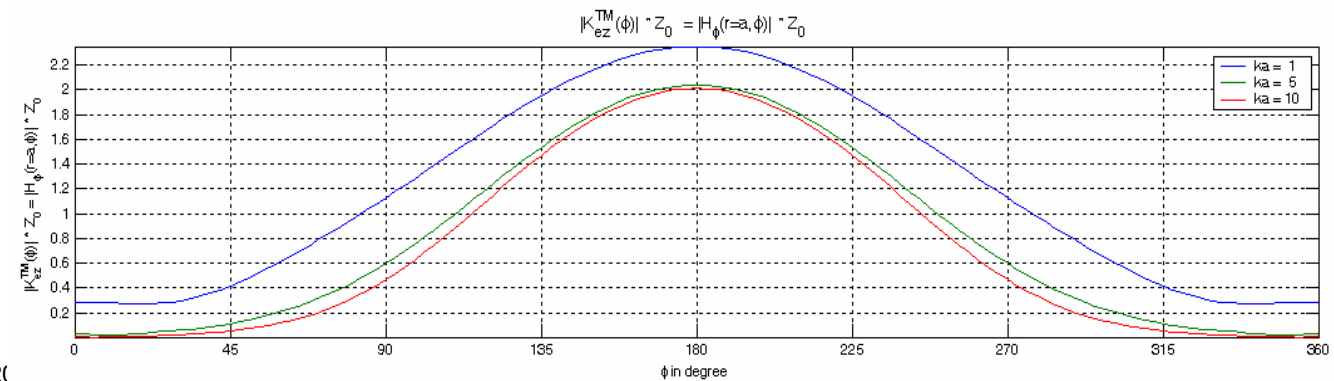
$$Z_0 \operatorname{Re}\{K_{ez}^{\text{TM}}(\varphi, \varphi_{\text{in}} = 0^\circ, \omega)\}$$



$$Z_0 \operatorname{Im}\{K_{ez}^{\text{TM}}(\varphi, \varphi_{\text{in}} = 0^\circ, \omega)\}$$

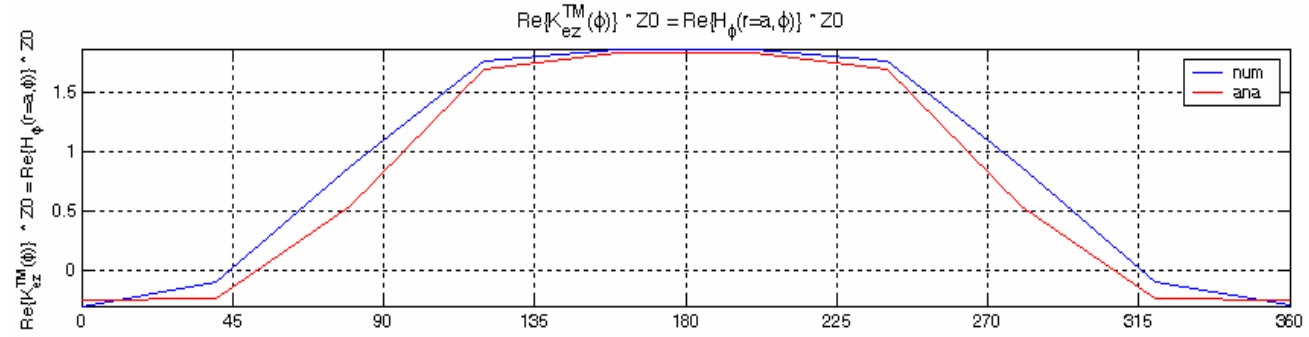


$$Z_0 |K_{ez}^{\text{TM}}(\varphi, \varphi_{\text{in}} = 0^\circ, \omega)|$$

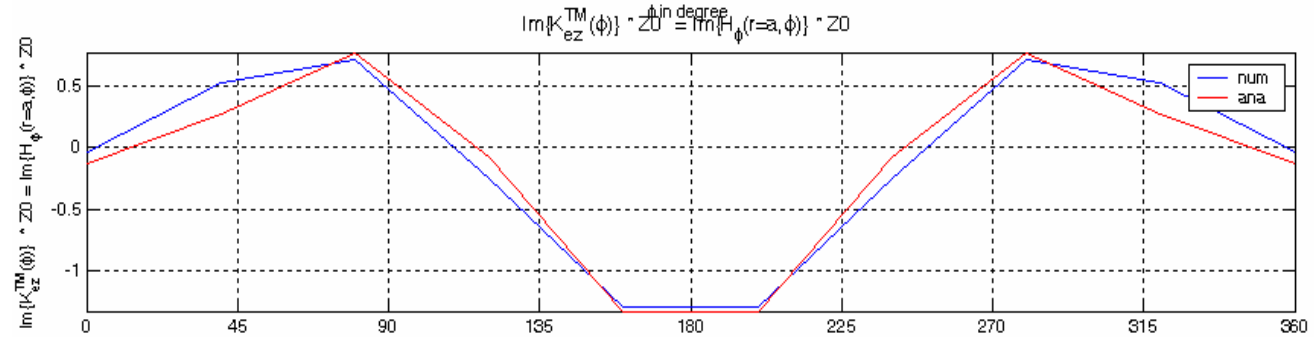


Induced Electric Surface Current for Different $ka = \{1, 5, 10\}$ and $N = 128$ / Induzierter elektrischer Flächenstrom für verschiedene $ka = \{1, 5, 10\}$ und $N = 128$

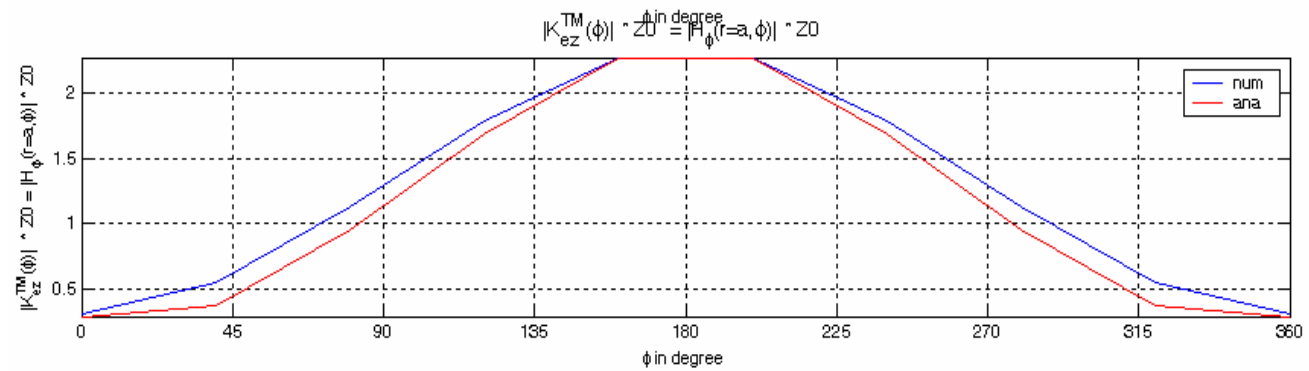
$$Z_0 \operatorname{Re}\{K_{ez}^{\text{TM}}(\varphi, \varphi_{\text{in}} = 0^\circ, \omega)\}$$



$$Z_0 \operatorname{Im}\{K_{ez}^{\text{TM}}(\varphi, \varphi_{\text{in}} = 0^\circ, \omega)\}$$



$$Z_0 |K_{ez}^{\text{TM}}(\varphi, \varphi_{\text{in}} = 0^\circ, \omega)|$$

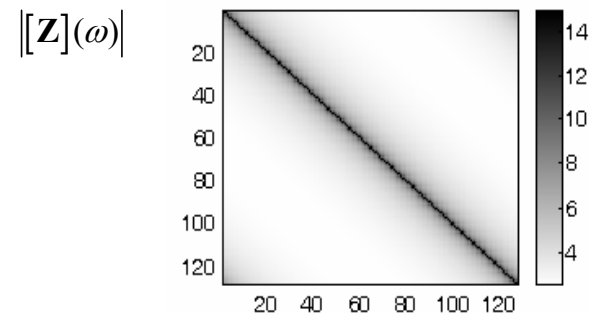
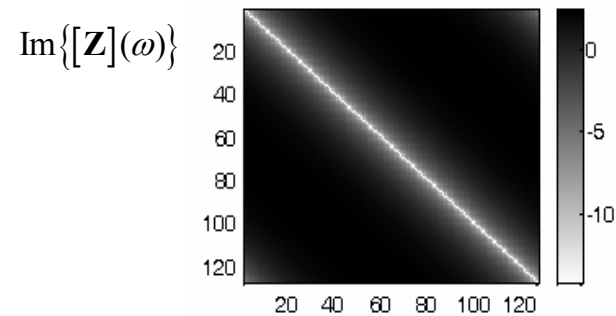
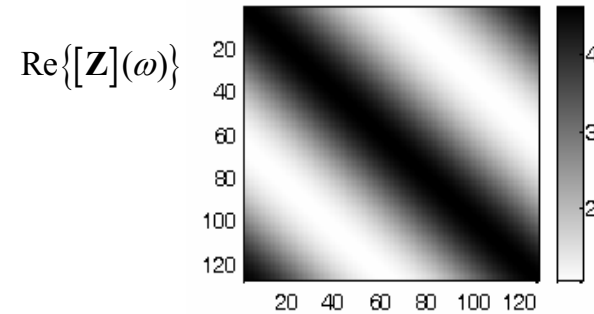


EM Scattering by a Perfectly Electrically Conducting Cylinder: EFIE Discretized in the 2-D TM Case with Pulse Basis and Delta Testing Functions / EM-Streuung an einem ideal elektrisch leitendem Zylinder: EFIE diskretisiert im 2D-TM-Fall mit Impuls-Basisfunktionen und Delta-Testfunktionen

Elements of the Impedance Matrix /
Elemente der Impedanzmatrix

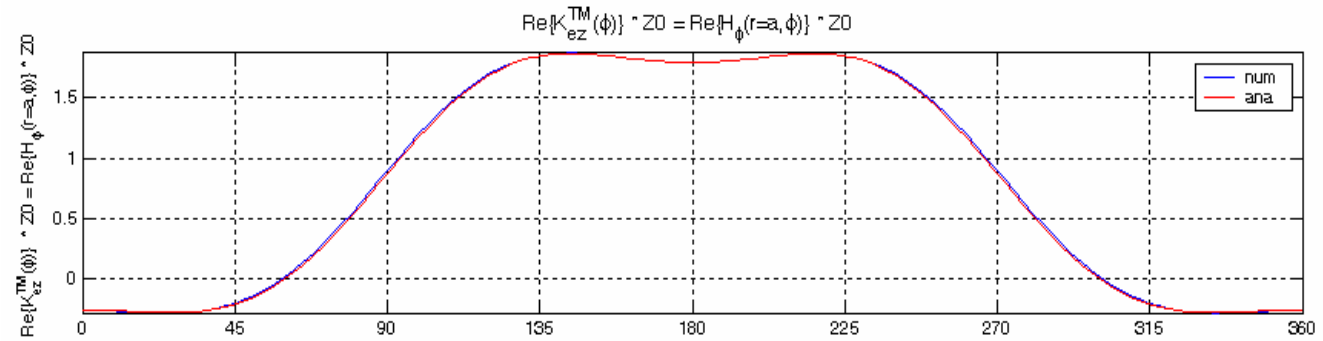
$$Z_{mn}(\omega) = \frac{\omega\mu_0}{4} \Delta^{(n)} \begin{cases} 1 + j\frac{2}{\pi} \left[\ln\left(\frac{k}{4} \Delta^{(n)}\right) + \gamma - 1 \right] & m = n \\ H_0^{(1)}(kr_{mn}) & m \neq n \end{cases}$$

$$ka = 1, \quad N = 128$$

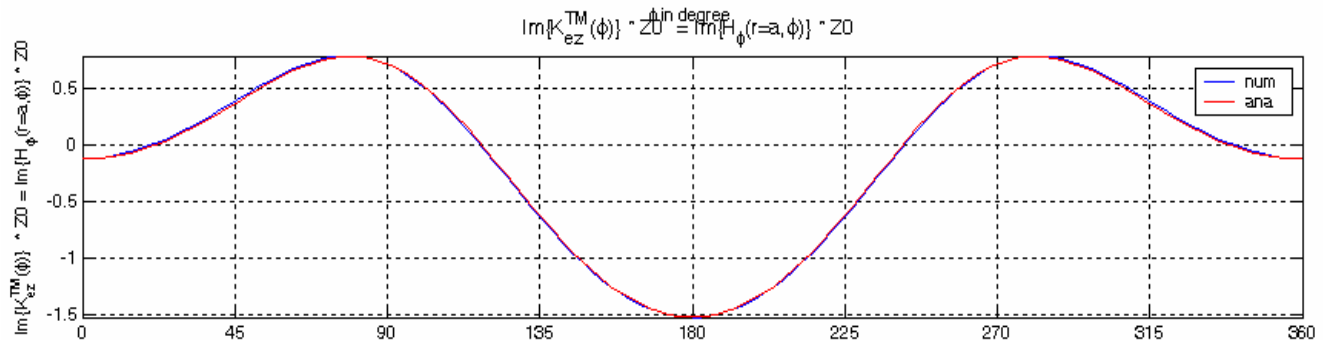


EM Scattering by a Circular PEC Cylinder – EFIE – 2-D TM Case – Results / EM-Streuung an einem kreisrunden IEL-Zylinder – EFIE – 2D-TM-Fall – Results

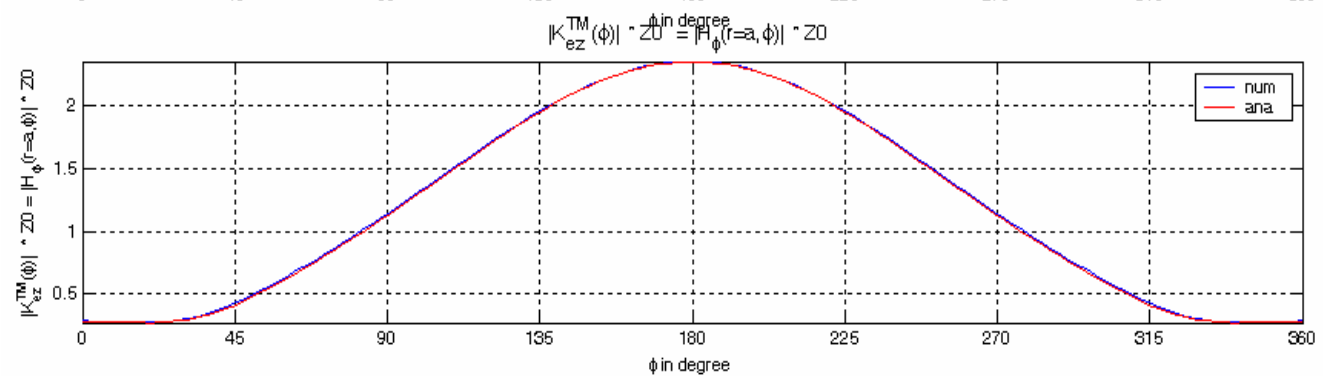
$$Z_0 \operatorname{Re}\{K_{ez}^{\text{TM}}(\varphi, \varphi_{\text{in}} = 0^\circ, \omega)\}$$



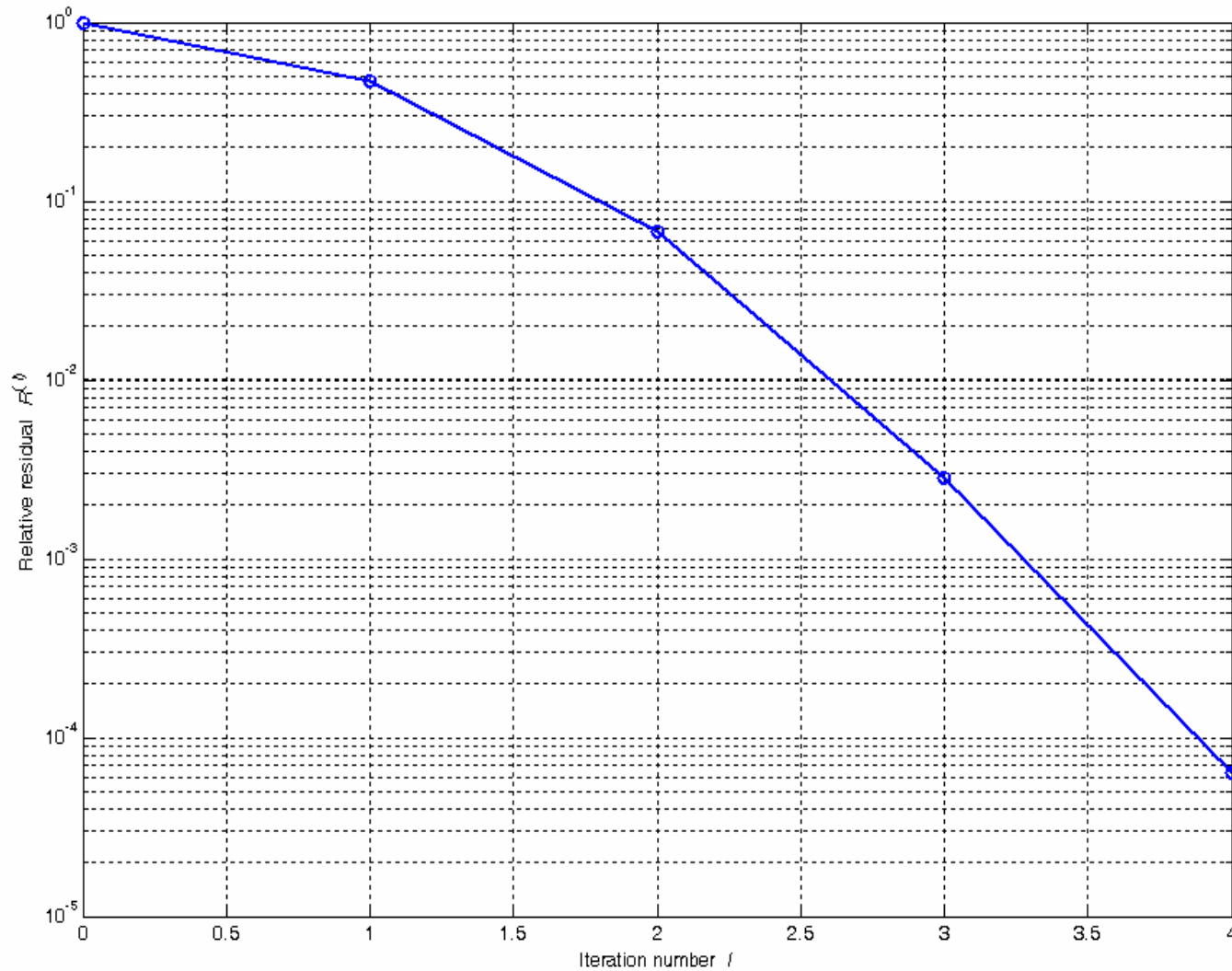
$$Z_0 \operatorname{Im}\{K_{ez}^{\text{TM}}(\varphi, \varphi_{\text{in}} = 0^\circ, \omega)\}$$



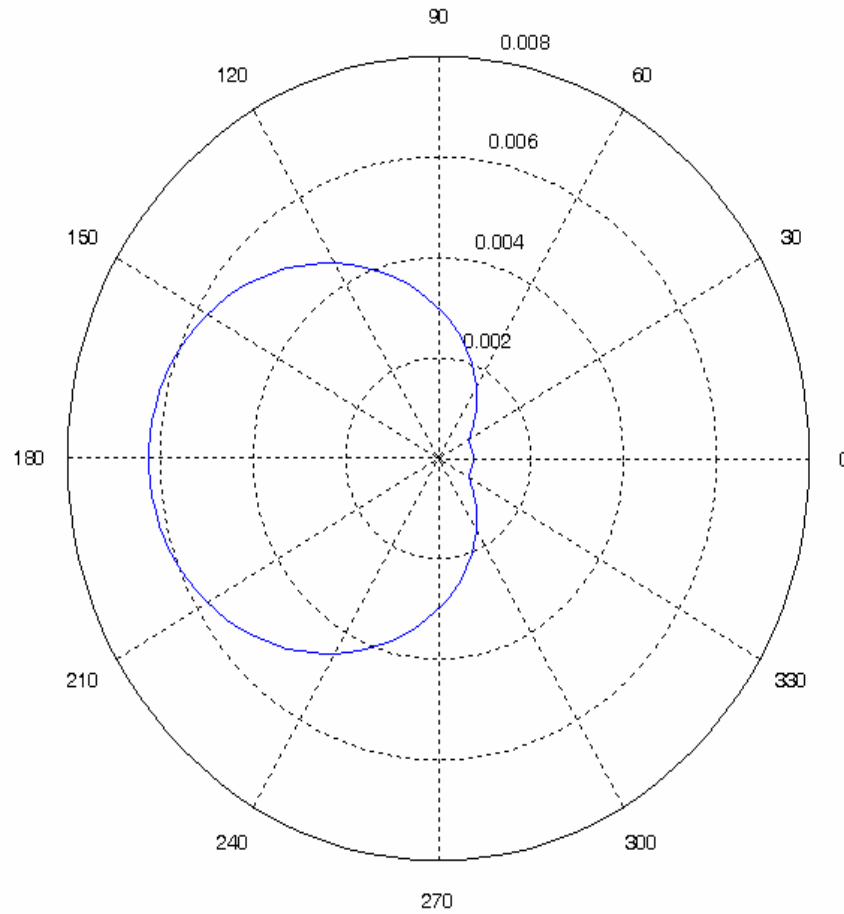
$$Z_0 |K_{ez}^{\text{TM}}(\varphi, \varphi_{\text{in}} = 0^\circ, \omega)|$$



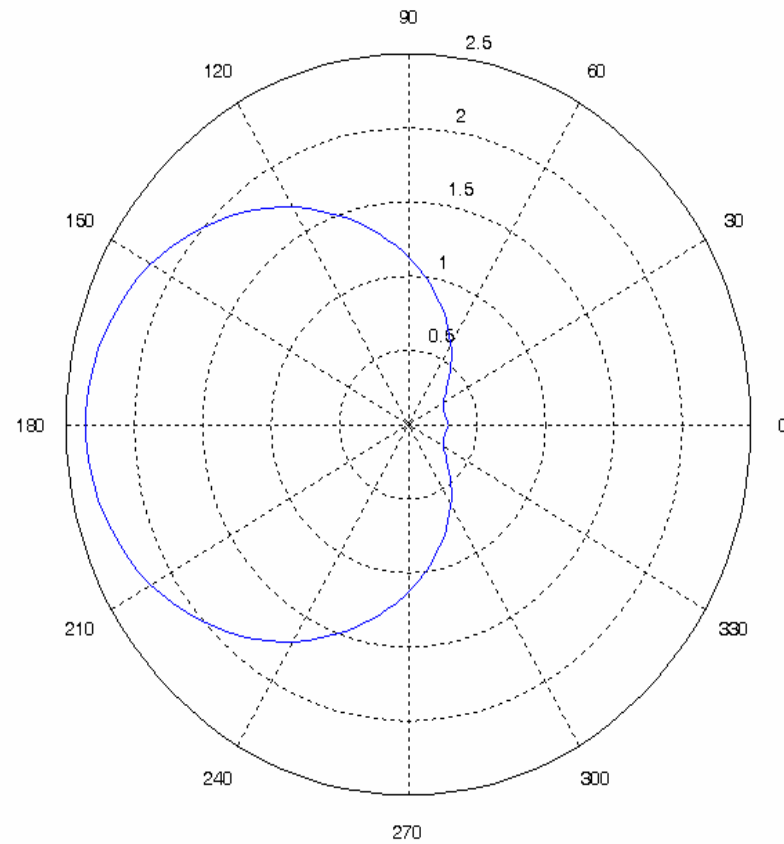
EM Scattering by a Circular PEC Cylinder – EFIE – 2-D TM Case – Results / EM-Streuung an einem kreisrunden IEL-Zylinder – EFIE – 2D-TM-Fall – Results



**EM Scattering by a Circular PEC Cylinder – EFIE – 2-D TM Case – Results /
EM-Streuung an einem kreisrunden IEL-Zylinder – EFIE – 2D-TM-Fall – Results**



**EM Scattering by a Circular PEC Cylinder – EFIE – 2-D TM Case – Results /
EM-Streuung an einem kreisrunden IEL-Zylinder – EFIE – 2D-TM-Fall – Results**



**End of 5th Lecture /
Ende der 5. Vorlesung**